

Prof. Dr.-Ing. Dr. h.c. Jürgen Becker (*becker@kit.edu*)

Dr.-Ing. Jens Becker (*jens.becker@kit.edu*)

Institut für Technik der Informationsverarbeitung (ITIV)

Communication Systems and Protocols

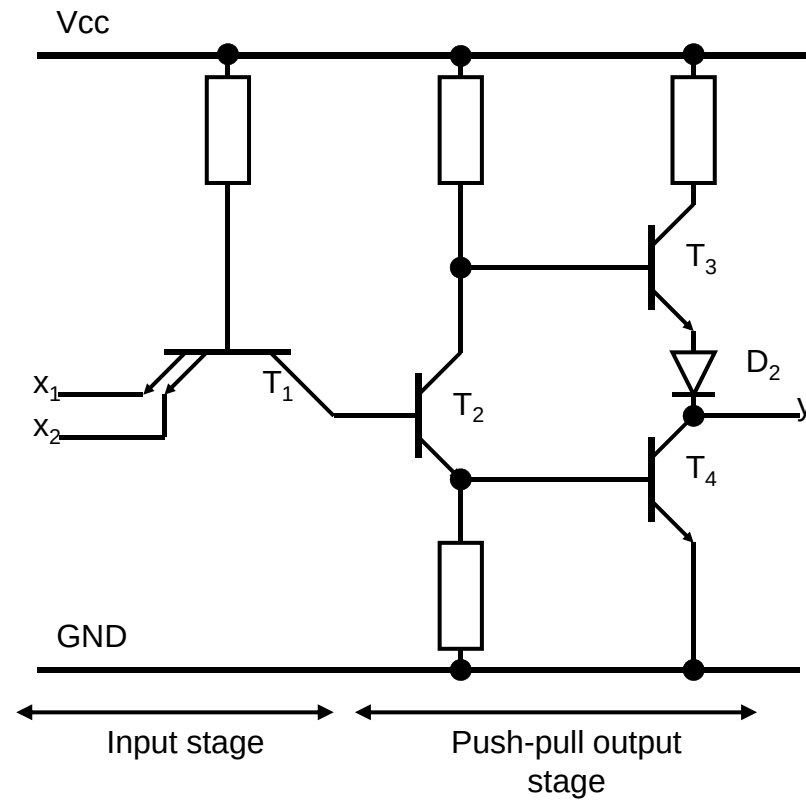
Session 3: Physical Principles

Clicker Session: Recapitulation

■ <https://arsnova.eu/mobile/#id/33969518>

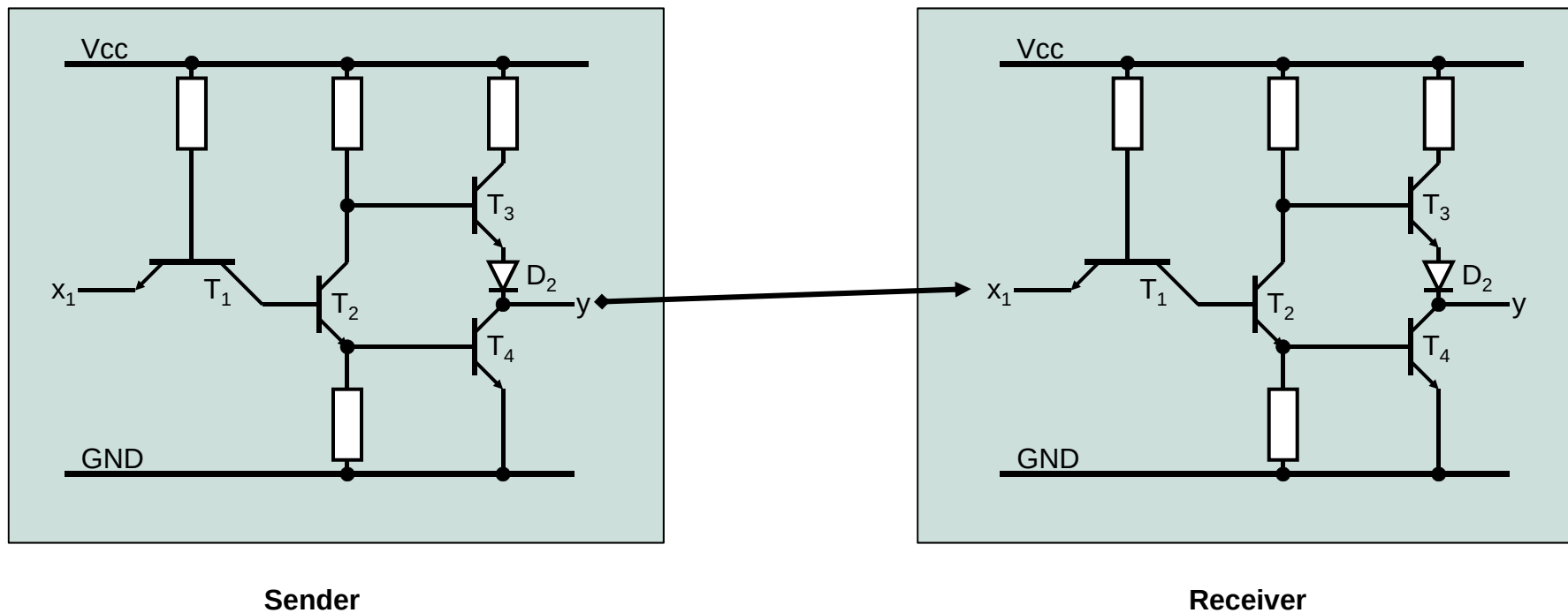


Recapitulation: TTL-Driver



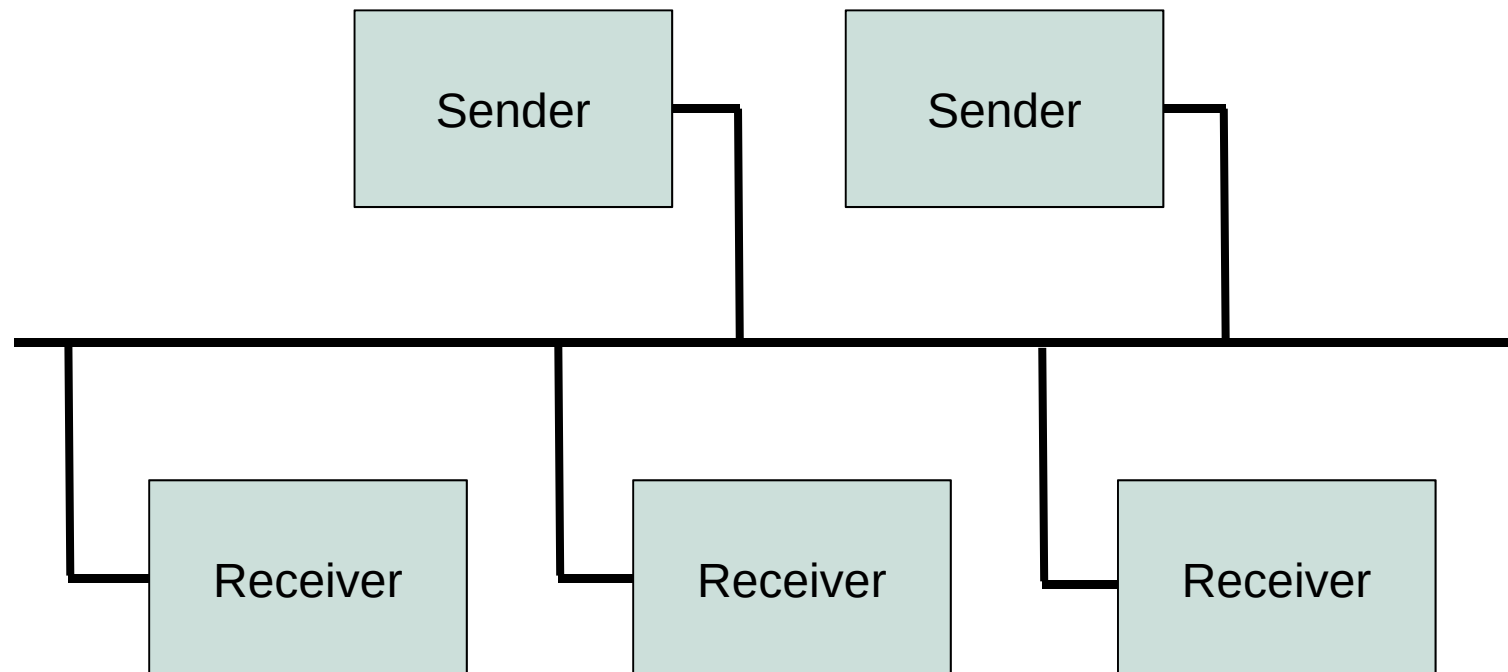
TTL basic gate (i.e. SN 7400)

Transmission Setup with TTL Technology



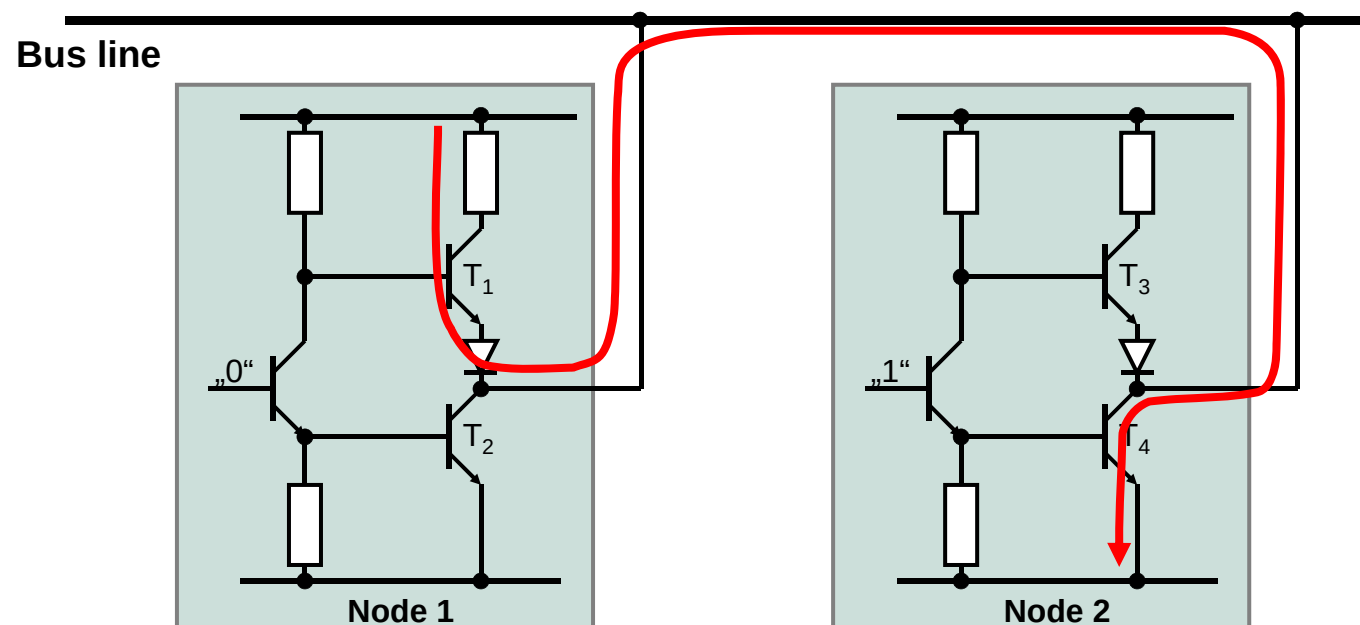
Idea of a Bus System

- Multiple Senders and receivers are connected together to the same channel for communication



Connection of multiple TTL drivers

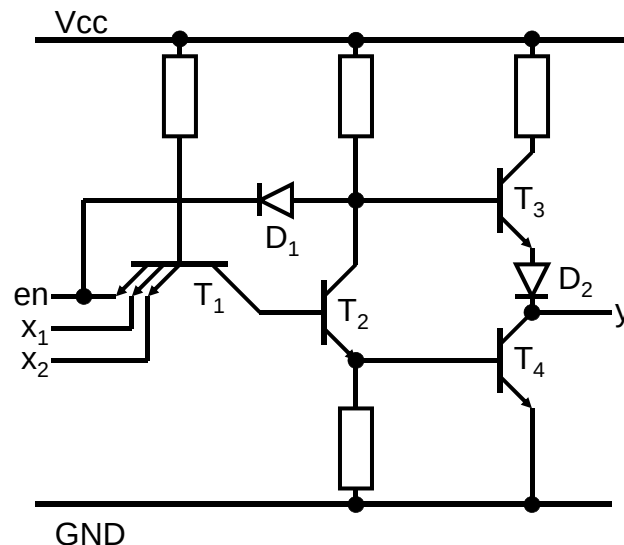
- Danger for short-circuit if the outputs of several TTL gates are connected to each other
 - If one gate is on HIGH level and another on LOW level a short-circuit current can flow via the open transistors (T1 and T4 in the figure)
- Modification of the bus coupling circuit is necessary when using driver on bus lines!



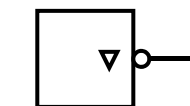
Solution 1: Tristate driver

- Parallel switching of several outputs is not possible with simple TTL gates --> danger of short-circuit
- By turning off unused output stages, several of such circuits can be connected to one bus
- Third state (Tristate): high-resistance (Z)
- In each case only one output is allowed to drive the bus, all others must show high-resistance

Example: NAND gate



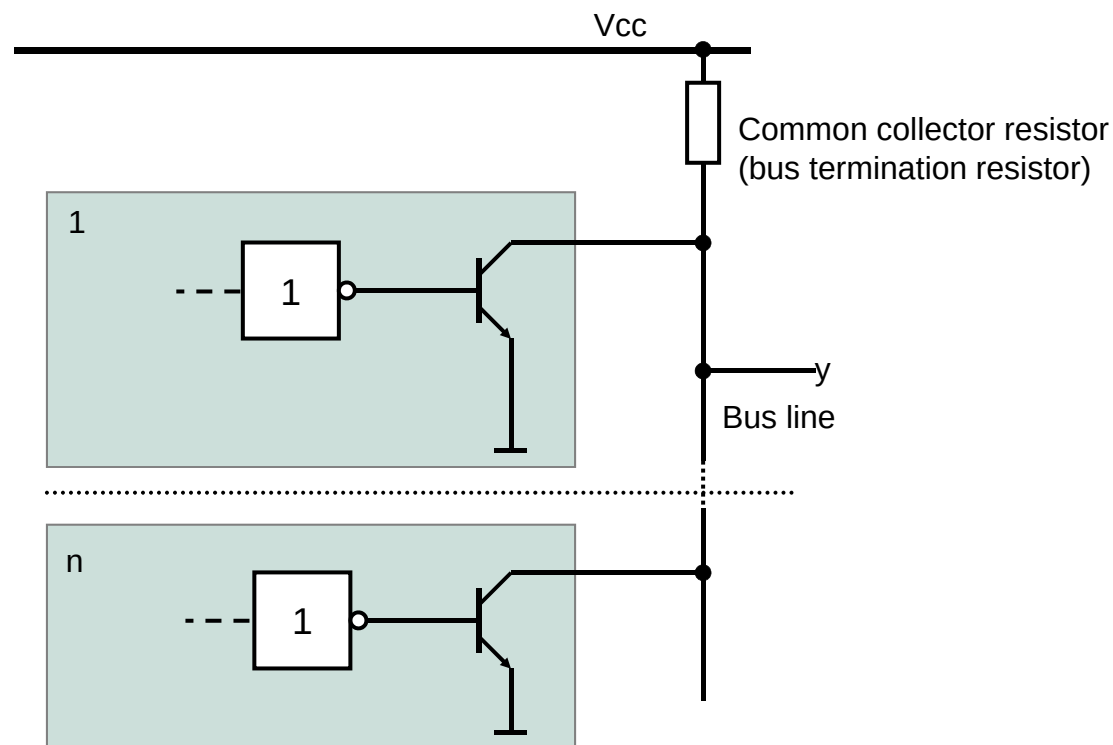
en	x ₁	x ₂	y
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	0
0	-	-	Z



Switch symbol
of a Tristate
output

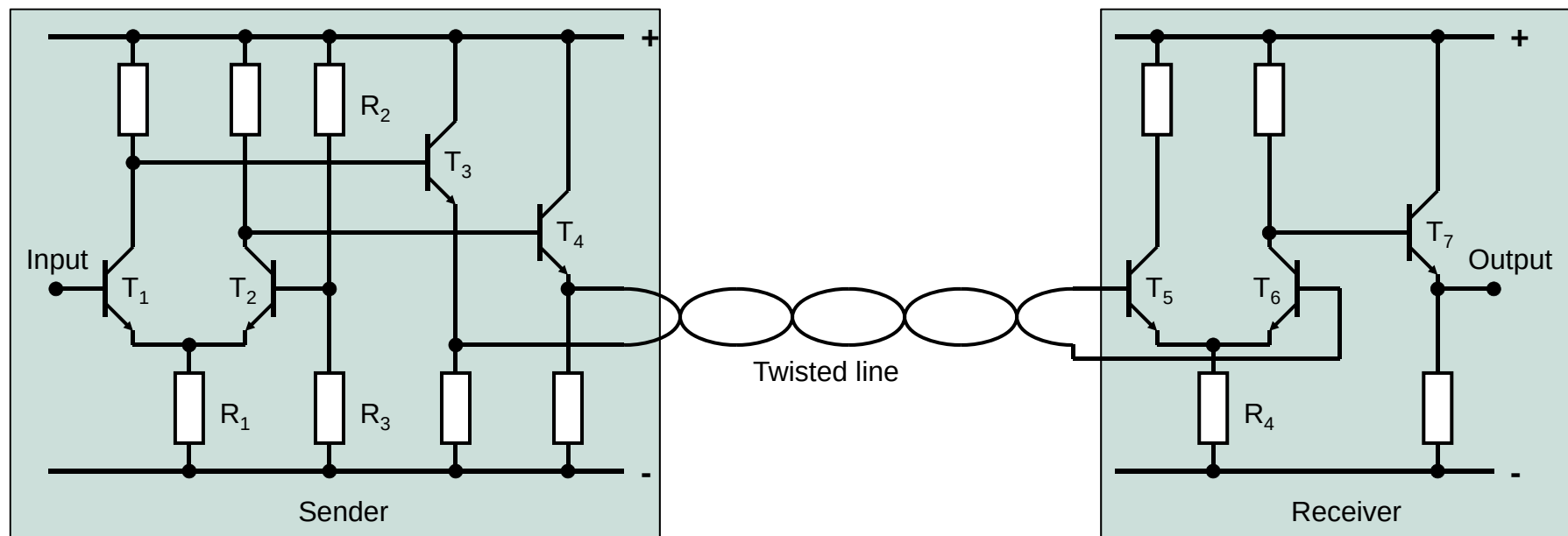
Solution 2: Open collector bus

- The collector of each of the output transistors remains unconnected
- All bus members share one collector resistor
- The output is only HIGH if all transistors cut off
 - low value (GND) on the bus line is dominant value



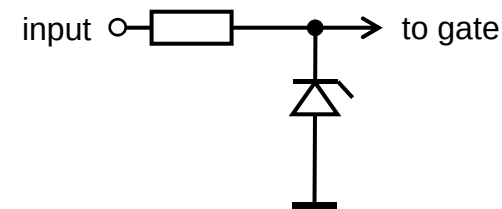
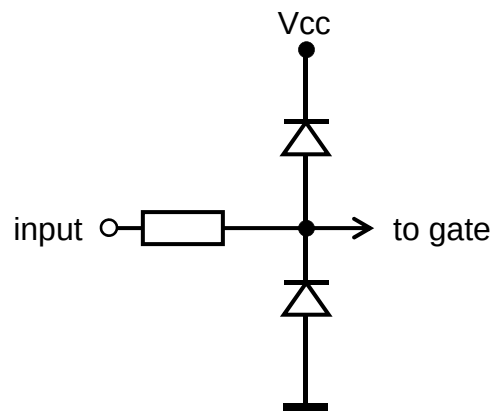
Example for differential signal generation

- Emitter Coupled Logic (ECL)
 - Higher speed since transistors do not go to saturation
 - With the help of the asymmetric input signal a differential signal for transmission is generated
 - The differential amplifier inside the receiver reconstructs the input signal

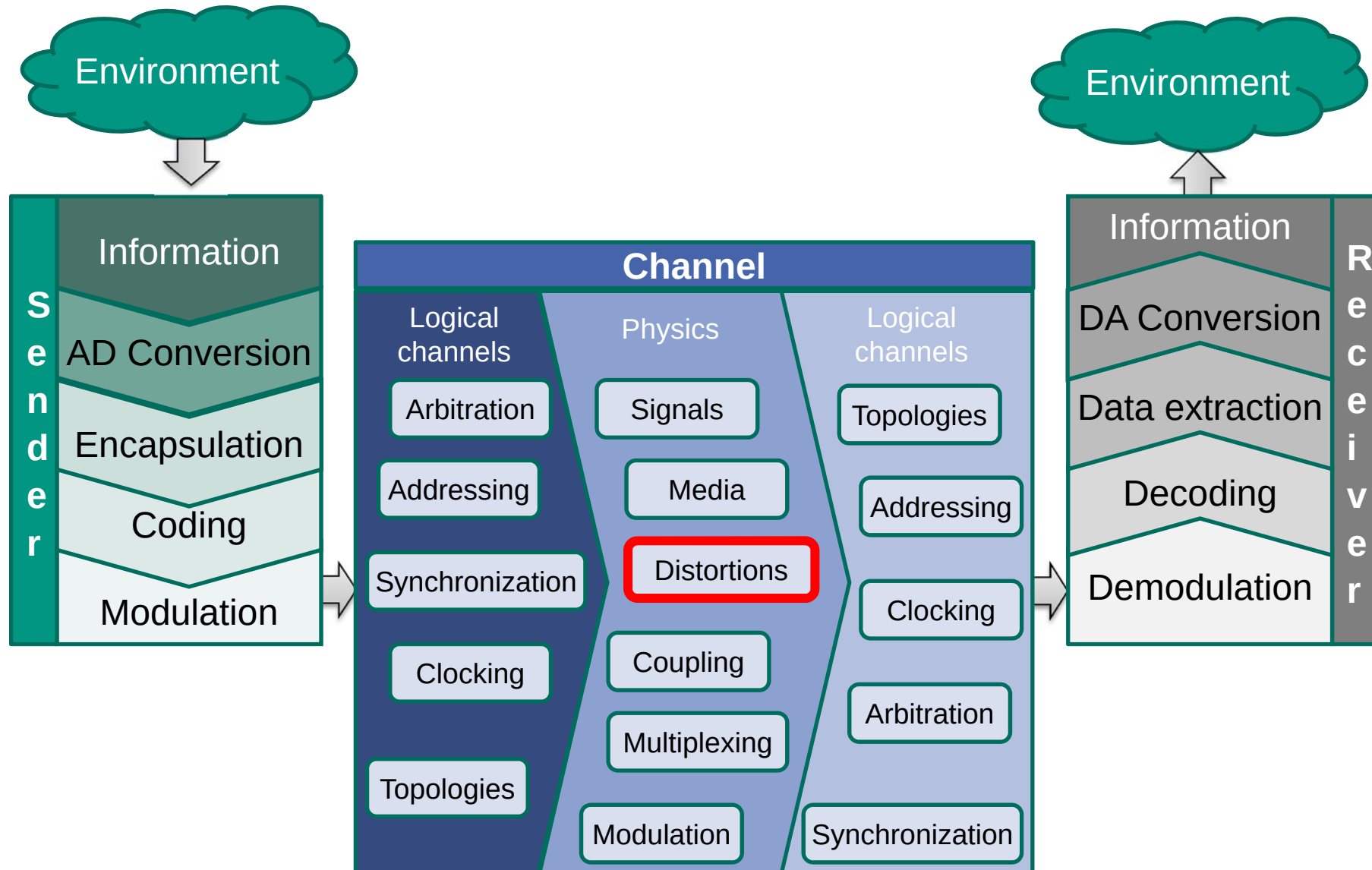


Input protection

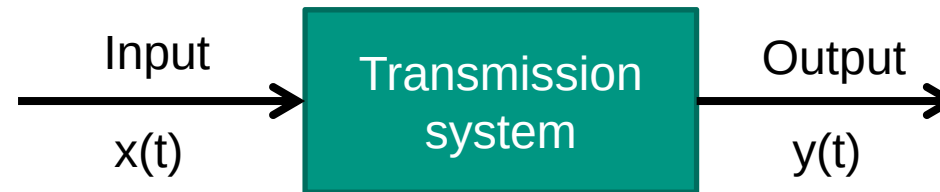
- High voltages at the input could damage the circuit
 - Limitation to $-0,7V \dots V_{cc} + 0,7V$ using two diodes
 - Alternate solution using Z-Diode
 - Current through diode is limited with additional resistor



Transmission System – Effects on electrical Signal Lines



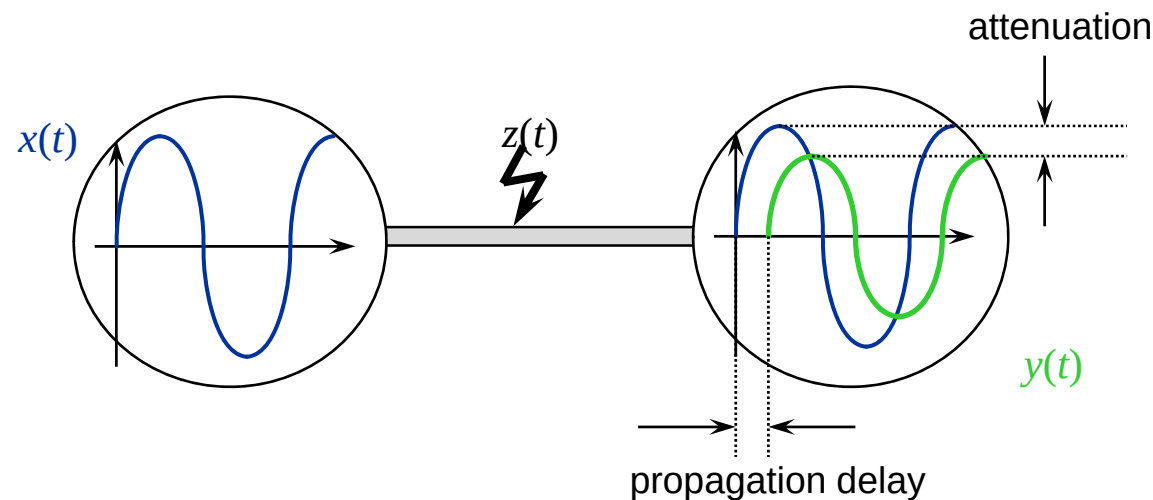
Signal Transmission over a Medium



■ with

- $x(t)$: Input signal
- $y(t)$: Output signal
- $z(t)$: Interference

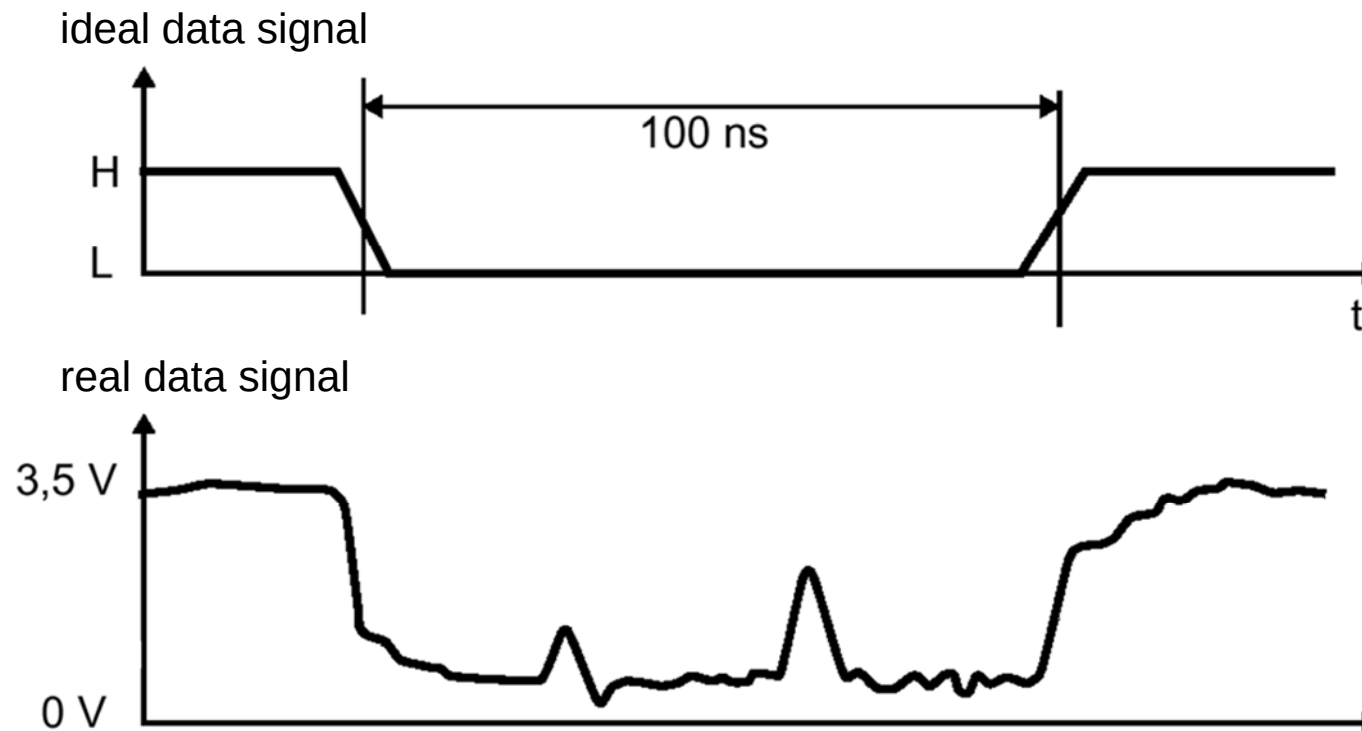
$$y(t) = F(x(t), z(t))$$



■ Interferences can be caused by external sources as well as by media itself

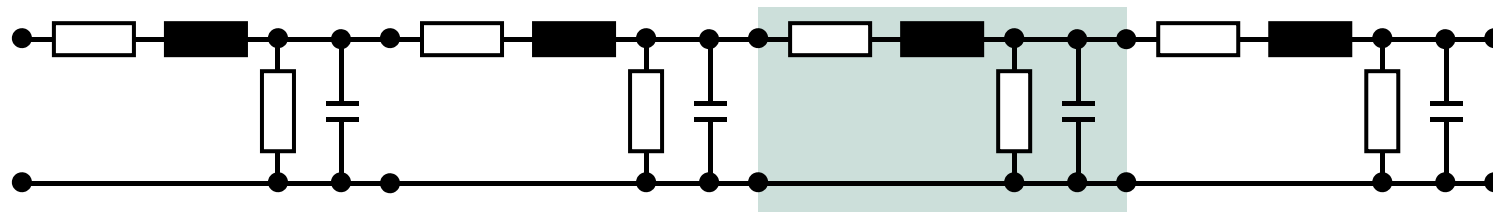
Possible signal distortions

- Reflections
- Attenuation
- Glitches
- Interference
- Noise
- Cross Talk



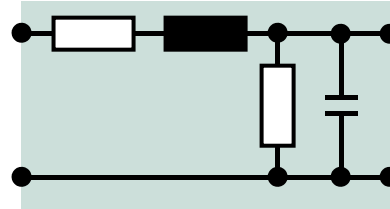
Model of Signal Lines

- Problem: When taking wave effects into account, signal lines can not be described by single elements any more
- Solution: Description using a distributed waveguide model → cascade of RLC quadripole



- Each segment represents an infinitesimally short segment of a transmission line

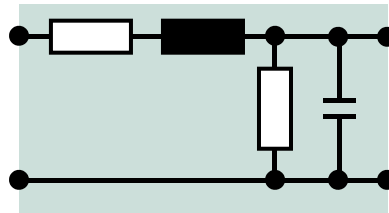
Components of the line model



- Distributed resistance R' , caused by resistance of the conductors is represented by a series resistor
- Distributed inductance L' , caused by the magnetic field around the wires is represented by a series inductor
- Distributed capacitance C' between the conductors is represented by a shunt capacitor
- Distributed conductance G' , caused by the dielectricum between the conductors is represented by a shunt resistor between the conductors

Idea: Derivation of the telegrapher's equation

- Boundary conditions: long, straight signal line, propagation in x-direction



- Using the laws of Kirchhoff, one can describe this system using the following equations:

$$\frac{\partial U(x, t)}{\partial x} = -L'(x) \frac{\partial I(x, t)}{\partial t} - R'(x) I(x, t)$$

$$\frac{\partial I(x, t)}{\partial x} = -G'(x) U(x, t) - C'(x) \frac{\partial U(x, t)}{\partial t}$$

Solution of the telegrapher's equation

- Under the assumption of constant distributed elements one can find:

$$U(x, t) = u_1 e^{i\omega t - \gamma x} + u_2 e^{i\omega t + \gamma x}$$

$$I(x, t) = i_1 e^{i\omega t - \gamma x} + i_2 e^{i\omega t + \gamma x}$$

$$\text{with } \gamma = \sqrt{(R' + i\omega L')(G' + i\omega C')}$$

- In the lossless case ($R'=0$, $G'=0$) the equations can be simplified:

$$U(x, t) = u_1 e^{i\omega t - \gamma x} + u_2 e^{i\omega t + \gamma x}$$

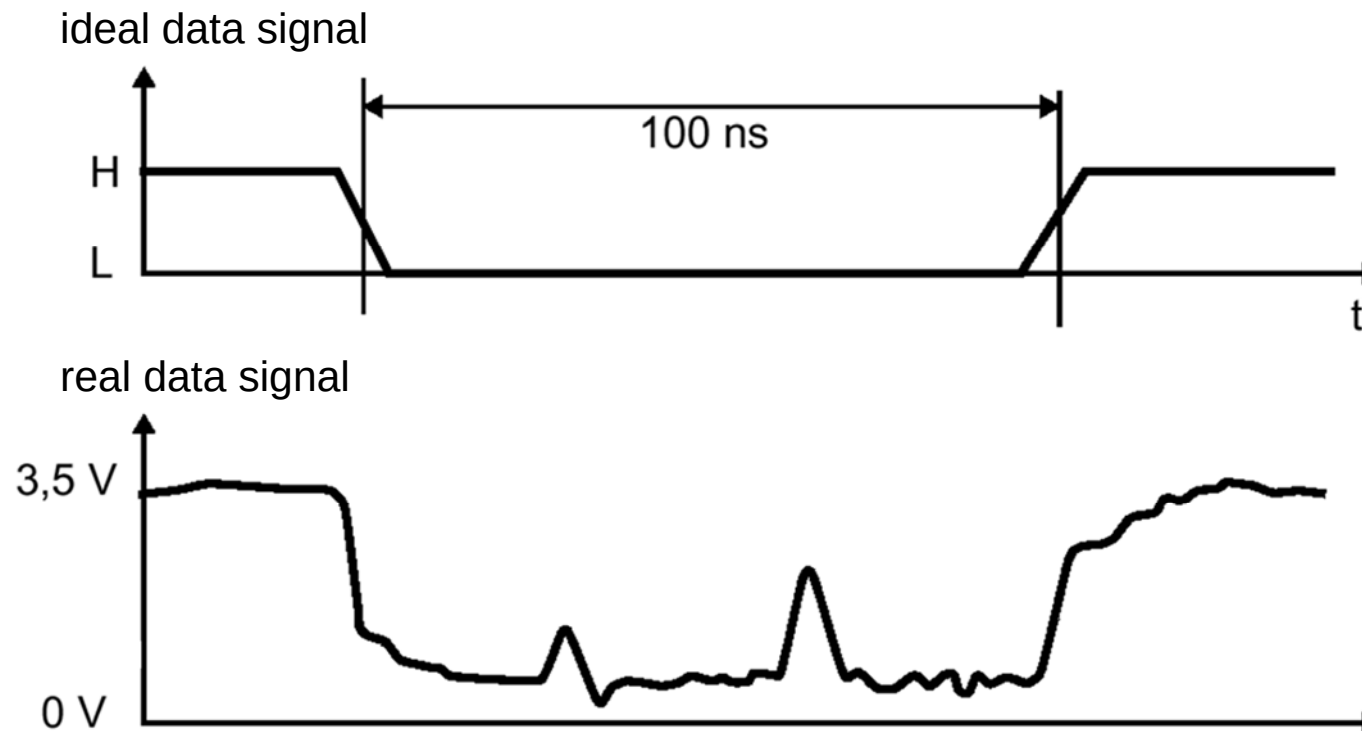
$$I(x, t) = \frac{u_1}{Z_0} e^{i\omega t - \gamma x} - \frac{u_2}{Z_0} e^{i\omega t + \gamma x}$$

$$\text{with } Z_0 = \sqrt{\frac{L'}{C'}}, \text{ that is called } \mathbf{\text{characteristic impedance}}$$

→ More details see lecture “Grundlagen der Hochfrequenztechnik”

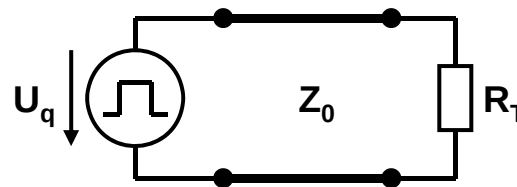
Possible signal distortions

- Reflections
- Attenuation
- Glitches
- Interference
- Noise
- Cross Talk



Reflection at the wire end

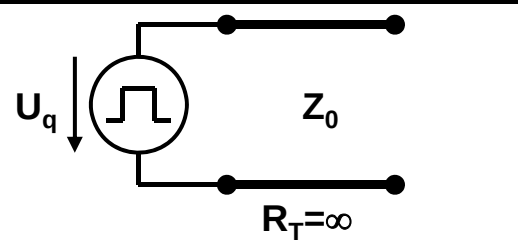
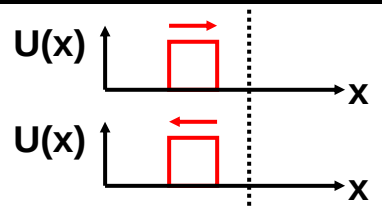
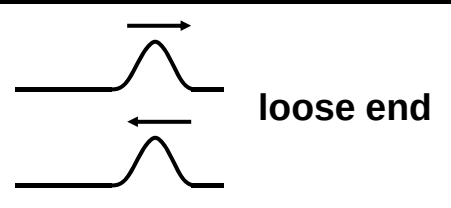
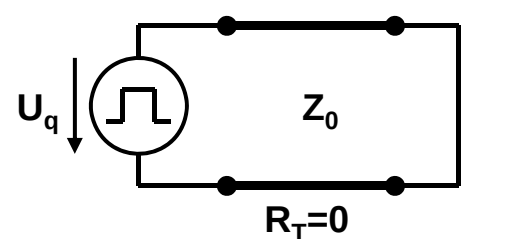
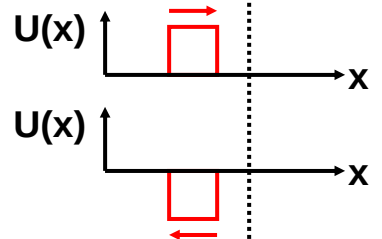
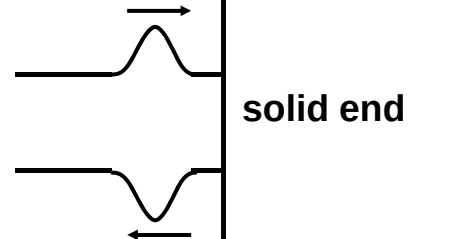
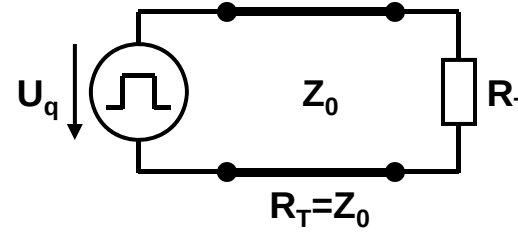
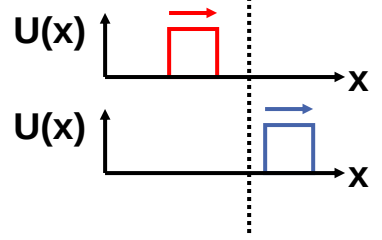
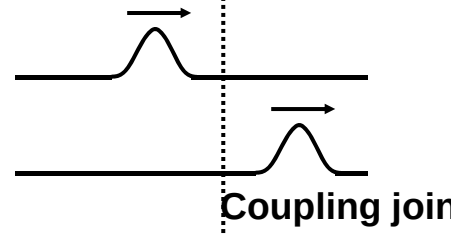
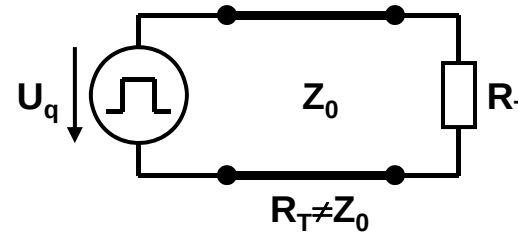
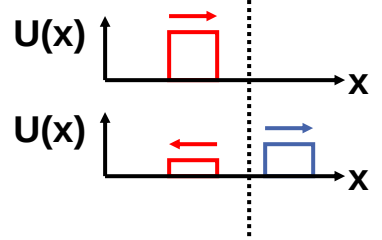
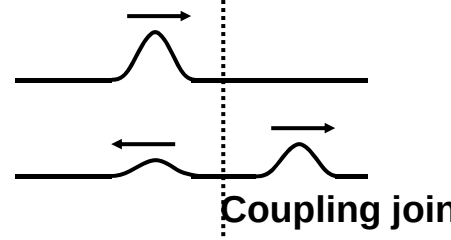
- Signals spread wavelike within long lines/wires
- There exists a forward and eventually a backward running wave that interfere on the wire
- The amplitude of the backward running wave depends on
 - The impedance of the line
 - The terminating impedance at the end of the line



- Definition of the reflection factor: $r = \frac{R_T - Z_0}{R_T + Z_0}$

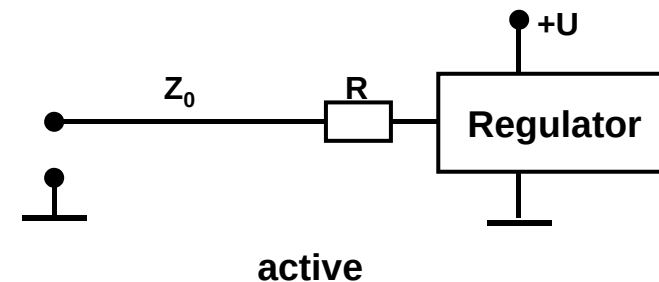
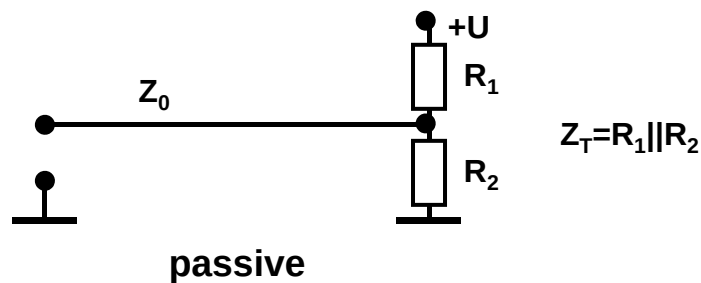
R_T : Terminal Resistance
 Z_0 : Characteristic impedance

Reflection on wires

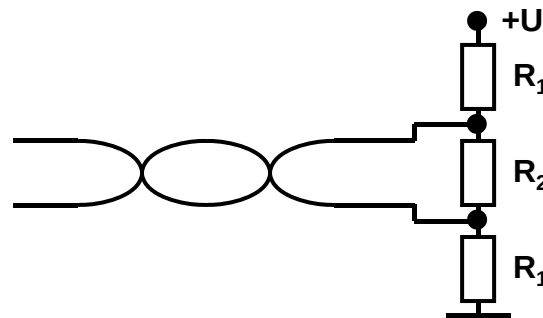
Line	Reflection factor r	Voltage curve	mechanical analogue
	$r=1$		 loose end
	$r=-1$		 solid end
	$r=0$		 Coupling joint
	$-1 < r < 1$		 Coupling joint

Line terminations

- Termination by using the characteristic wave resistance Z_0 to avoid reflections
- The utilization of several resistors switched in parallel is advantageous since usage of resistors of a relatively high value which provide for a lower DC charge allow for a approximation to the relatively low resistance of the line
- Asymmetric line

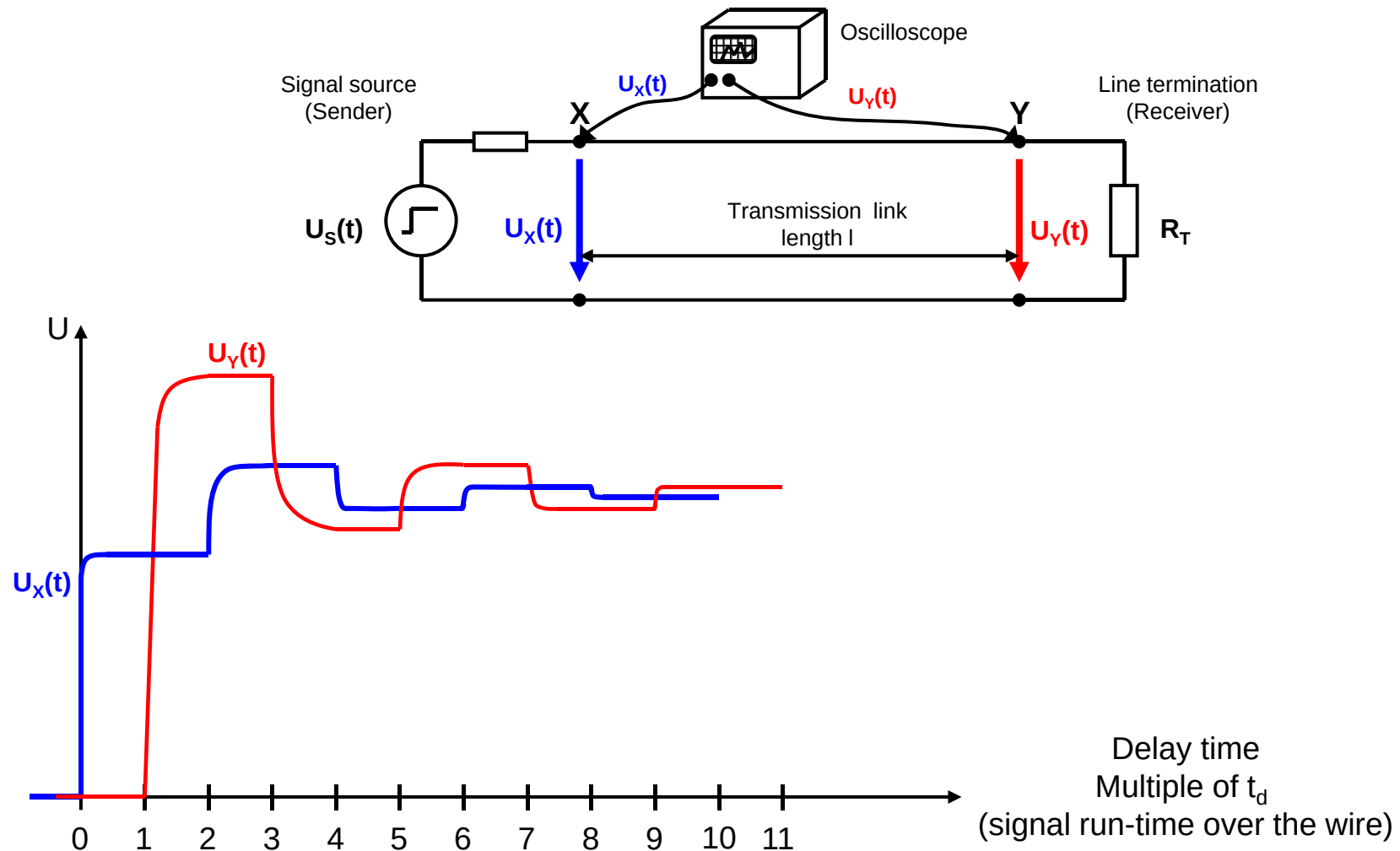


- Symmetric line



Outlook to exercises: Reflection on wires

- Measurement setup for the determination of the reflection on the lines:



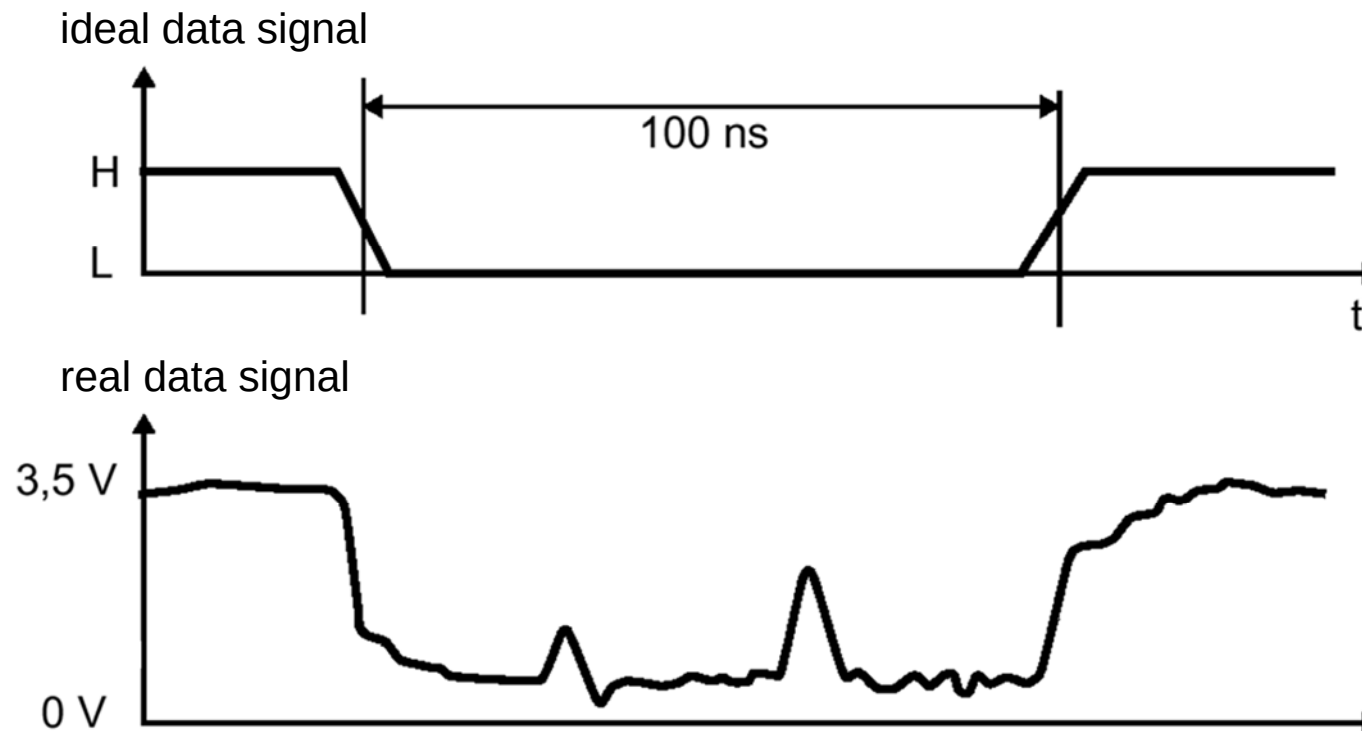
Clicker Session: Reflection on wires

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Possible signal distortions

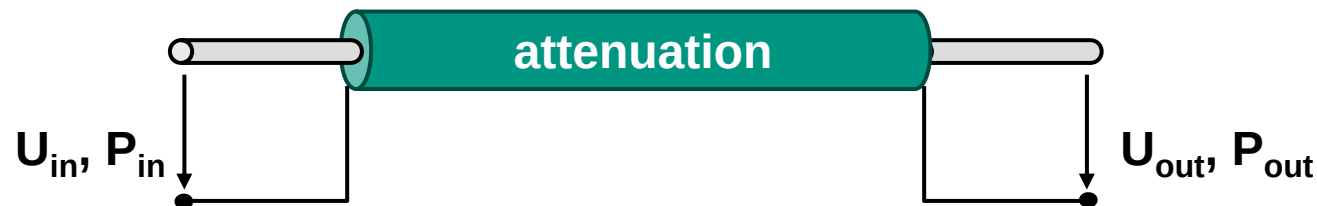
- Reflections
- **Attenuation**
- Glitches
- Interference
- Noise
- Cross Talk



Attenuation (media dependend)

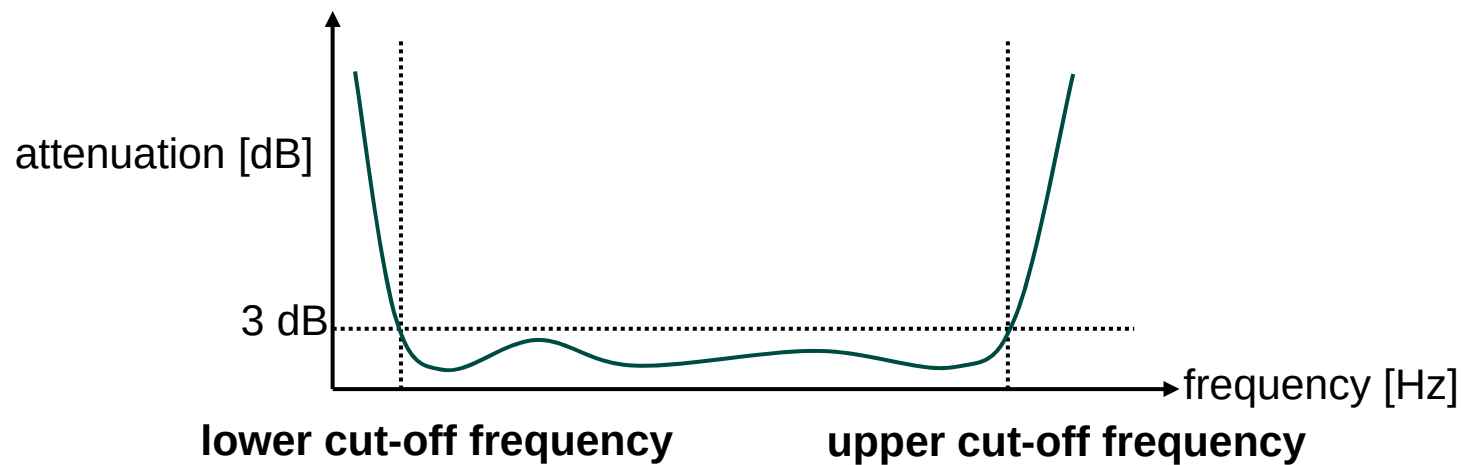
- Attenuation of signals is caused by:
 - Ohmic loss
 - Dielectric loss
 - Radiation loss
 - Reflections

- Attenuation $G = 10 \log(P_{in} / P_{out}) = 20 \log(U_{in} / U_{out}) [dB]$
 - additive properties of single cable sections
 - Depends on length/distance
 - Depends on frequency of transmitted signal



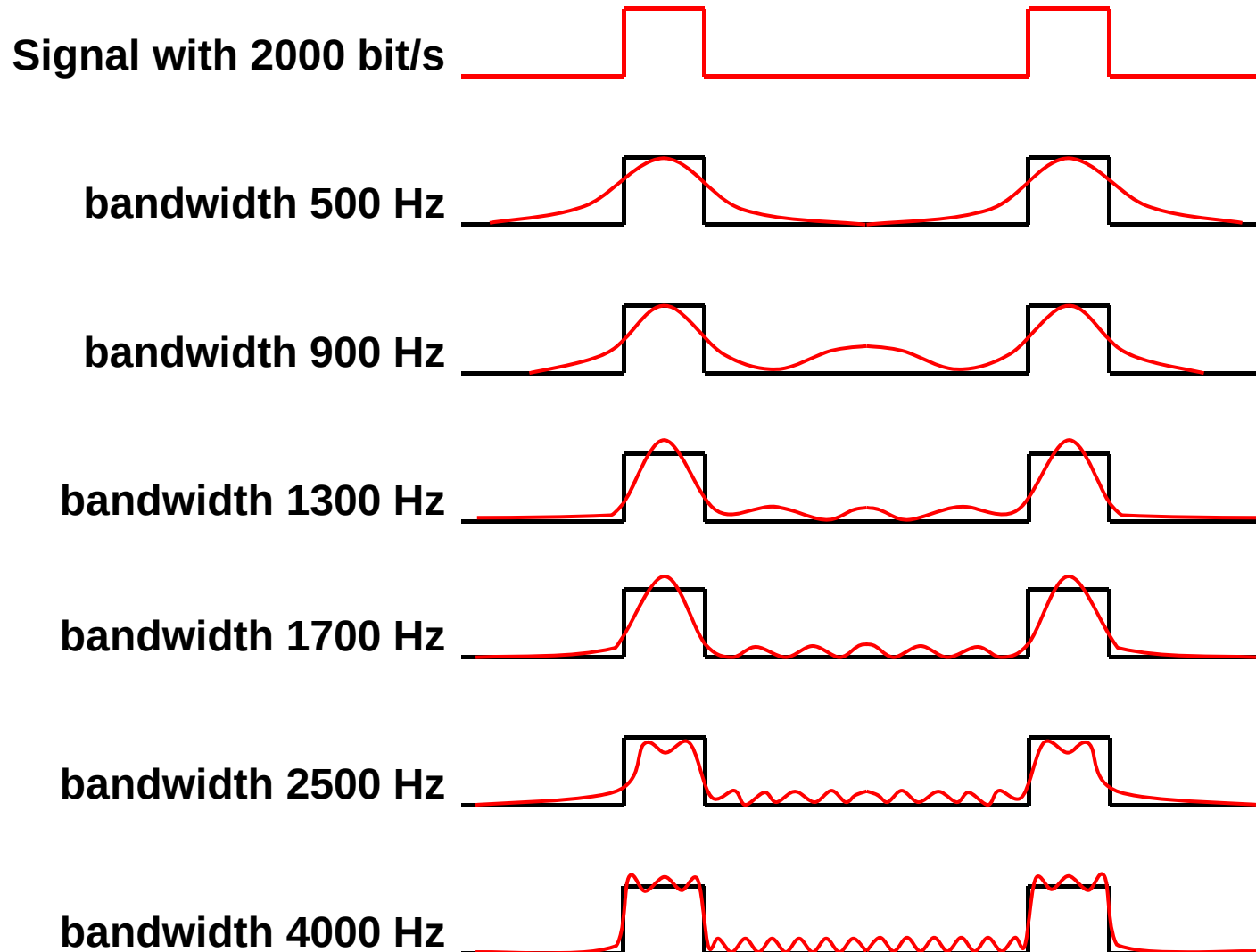
Bandwidth

- Cut Off frequency
 - Frequency at which the signal amplitude has dropped by 3dB compared to the output value
- Interval between upper and lower cut-off frequency
 - Between the frequencies almost uniform attenuation



- Limits not ideal
 - No vertical signal edges
 - Frequencies are attenuated at different degrees

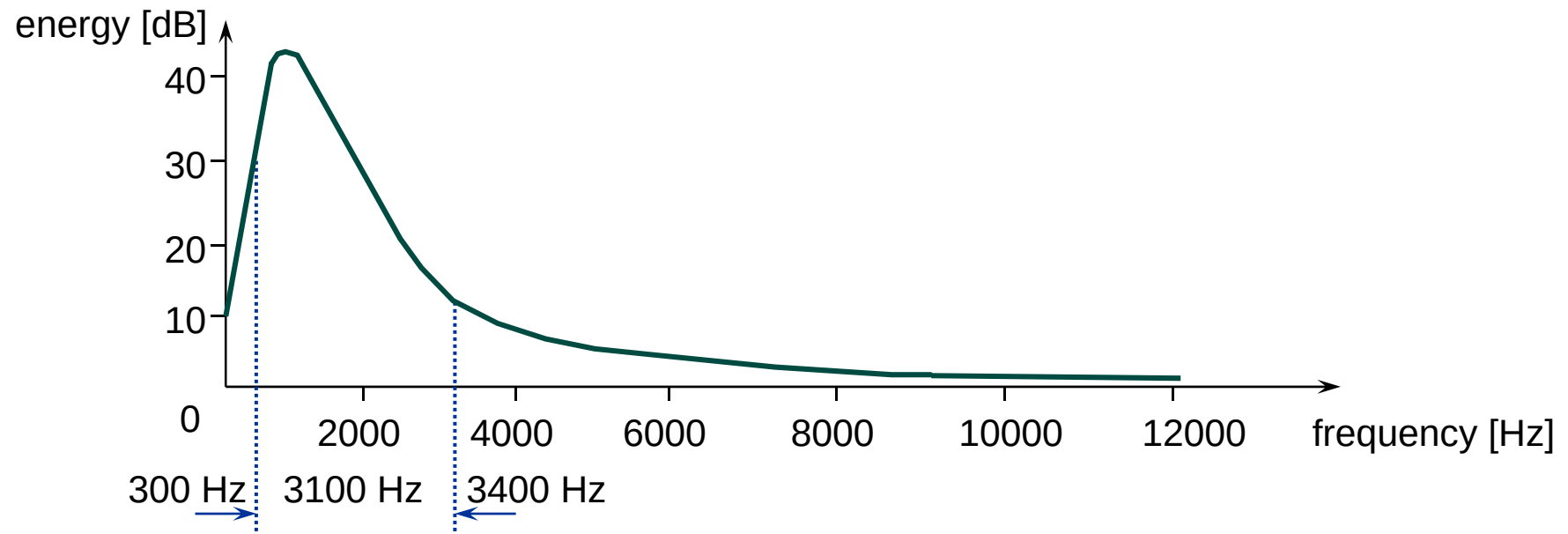
Influence of the Bandwidth on a digital Signal



source: University Ulm

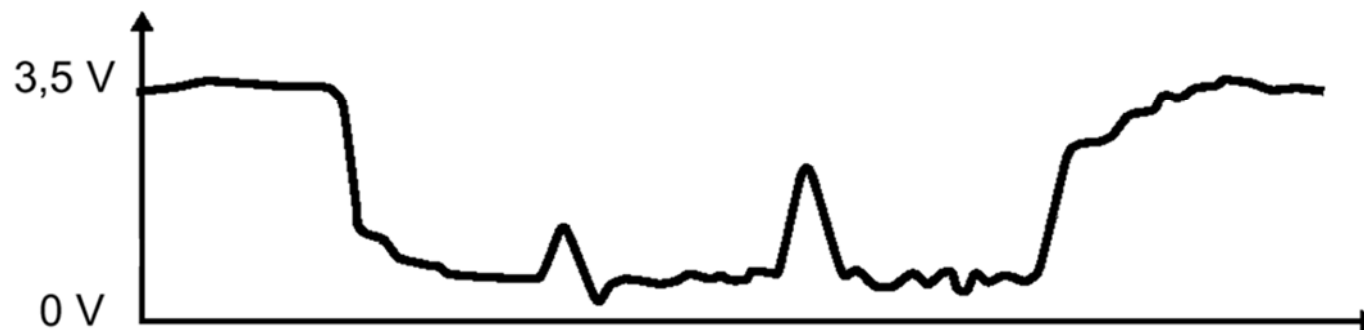
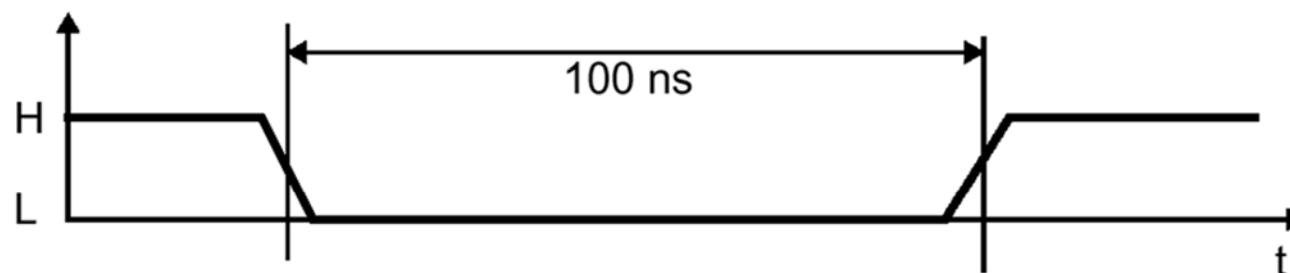
Example: Frequency Spectrum of a Signal

- Band limited signal: signals can comprise a “naturally” limited frequency spectrum or can be limited to a section of a spectrum through technical means (bandwidth)
 - Example: ITU-Standard telephone channel
Frequency spectrum of the human voice

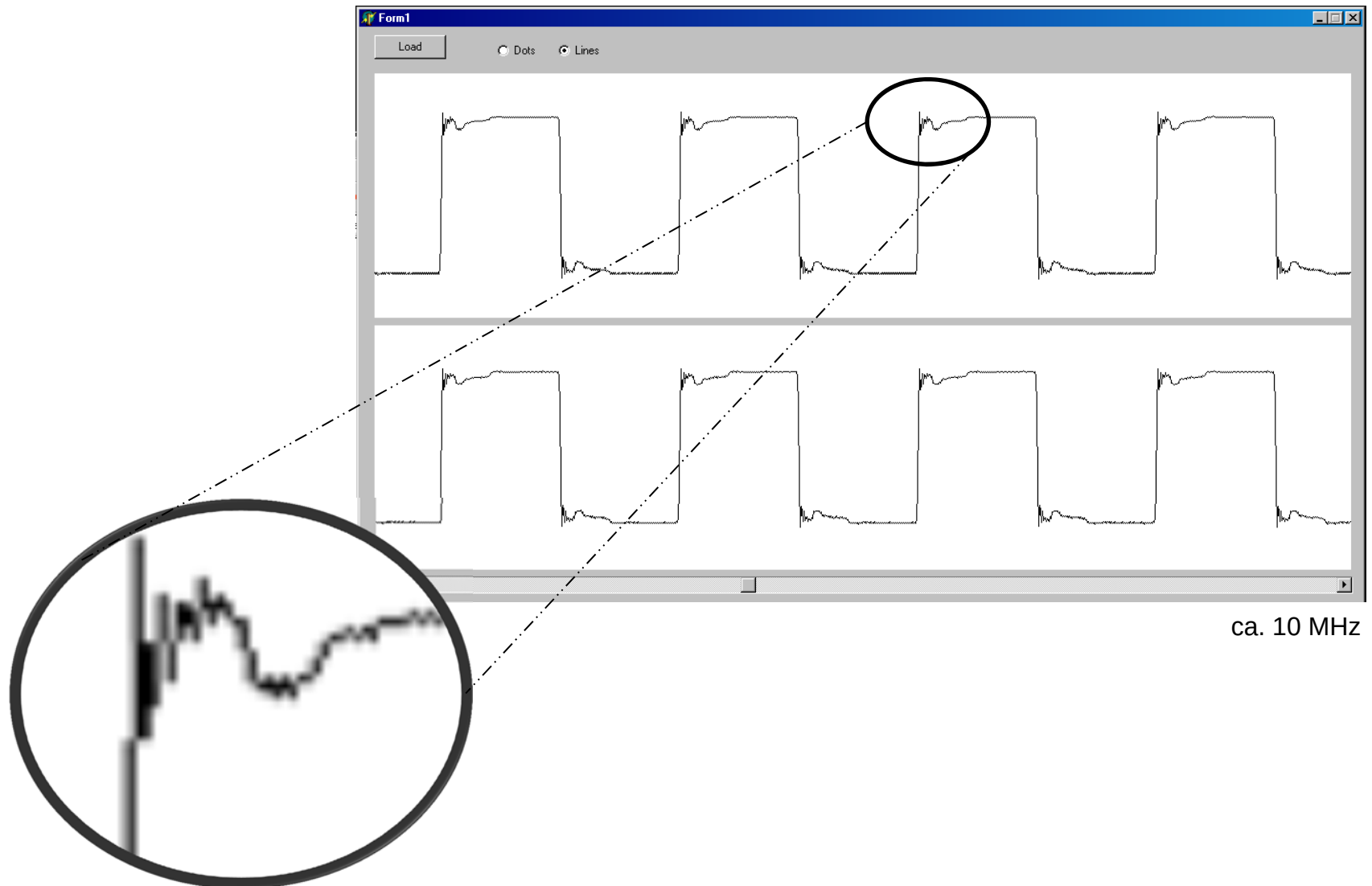


Possible Signal Distortions

- Reflections
- Attenuation
- Glitches
- Interference
- Noise
- Cross Talk



Real Signal

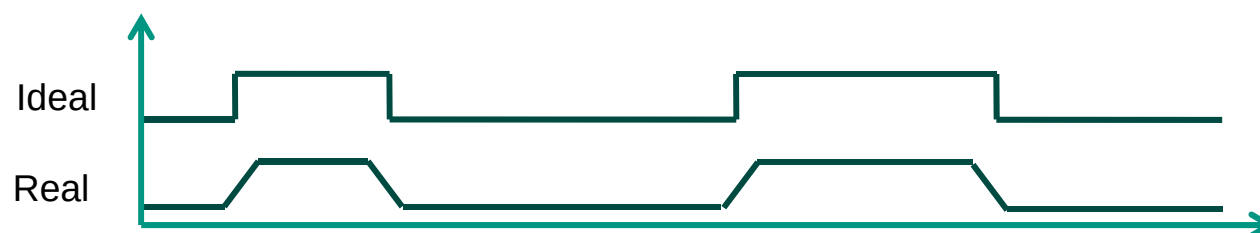


ca. 10 MHz

Excursus: Representation of Signals

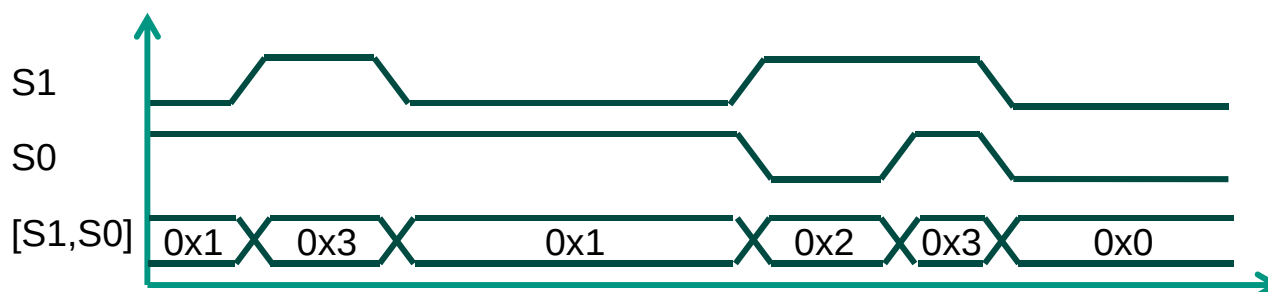
■ Ideal vs. realistic edges

- Raise and fall time is approximated by slanted edges



■ Combination of individual lines (busses)

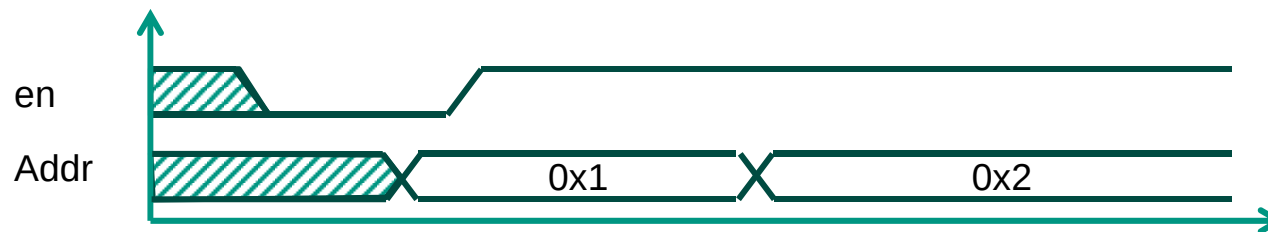
- Signal is drawn as a ribbon with the combined value as text
- Every change on one of the signals results in a change of the bus



Excursus: Representation of Signals II

■ Undefined value

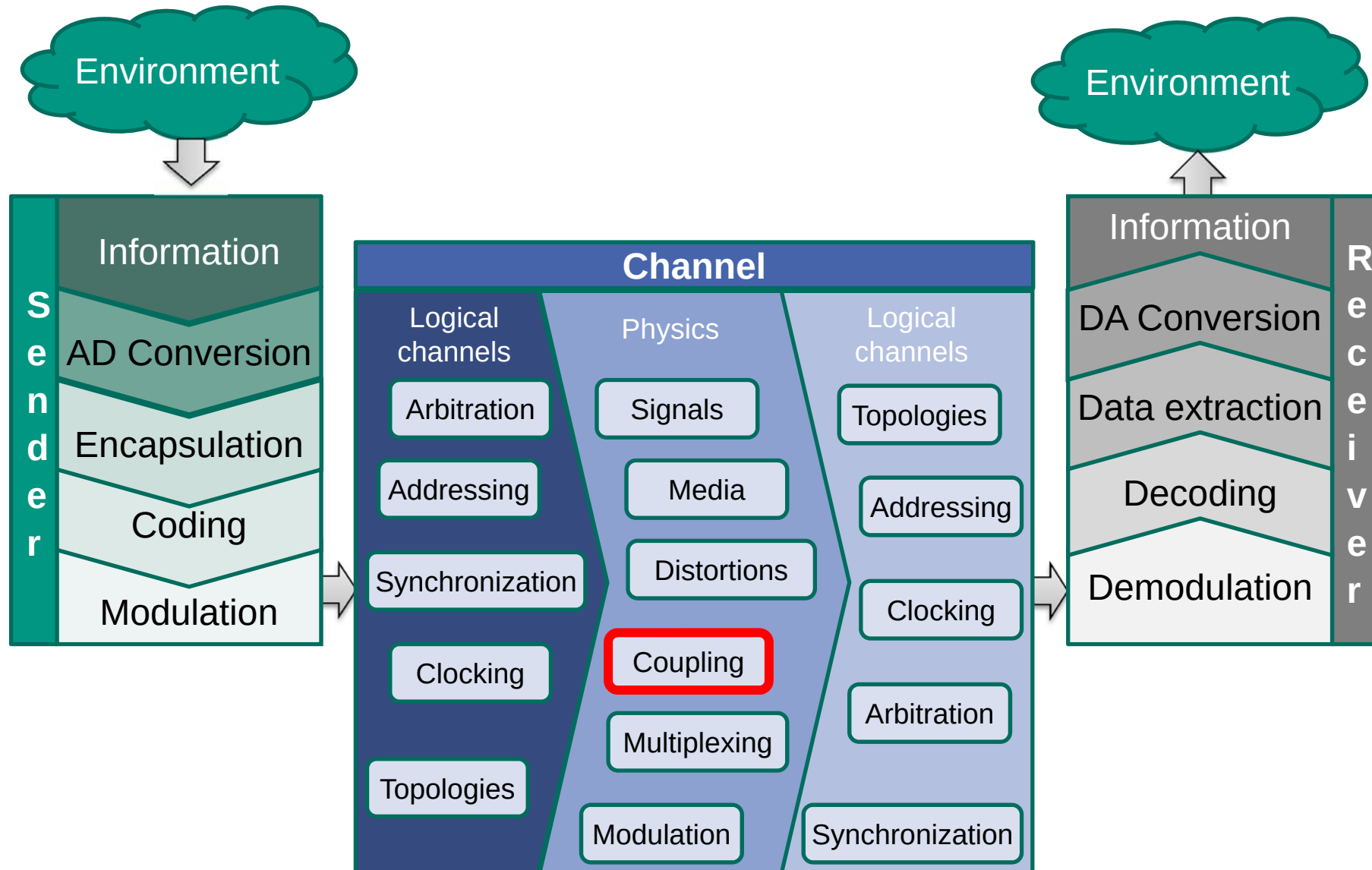
- In specifications, a signal might not be determined for every point in time
- Undefined region is depicted as shaded area



■ Inverted Logic (LOW active)

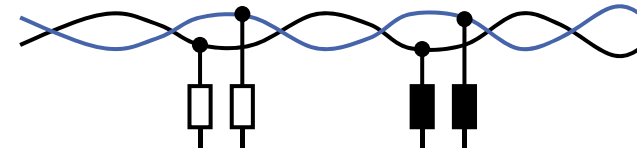
- Signals with inverted logic can be marked as follows:
 - $\bar{n}\text{Signal}$
 - $^*\text{Signal}$
 - Signal

Transmission System – Media Coupling

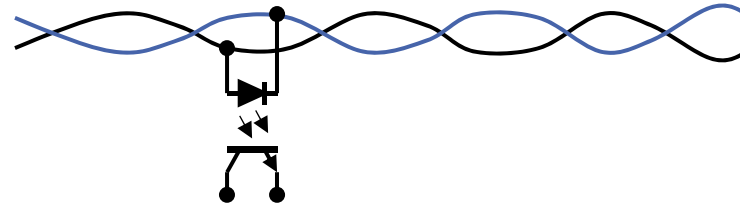


Media coupling

- Galvanic coupling
 - Ohmic resistors for protection
 - Series inductance when having an interfering source voltage



- Galvanically separated coupling
 - Avoidance of disturbances by interference on great distances

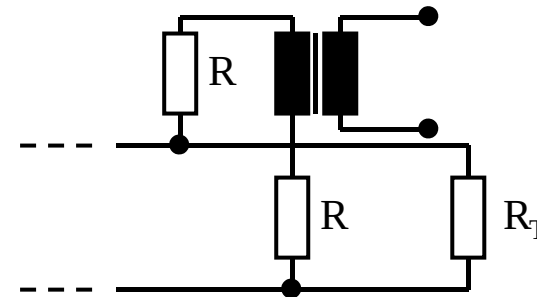
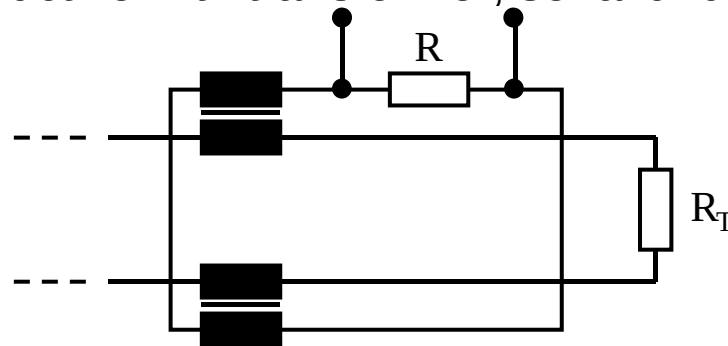


- Optic coupling by using optoelectronic couplers
 - Luminous intensity is quasi-statically controlled
→no lower cut-off frequency

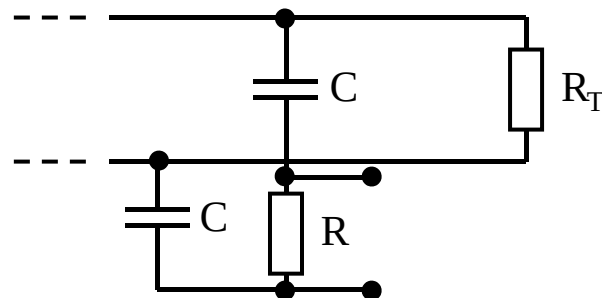
Media coupling

■ Galvanically separated

■ Inductive with transformer, serial und parallel



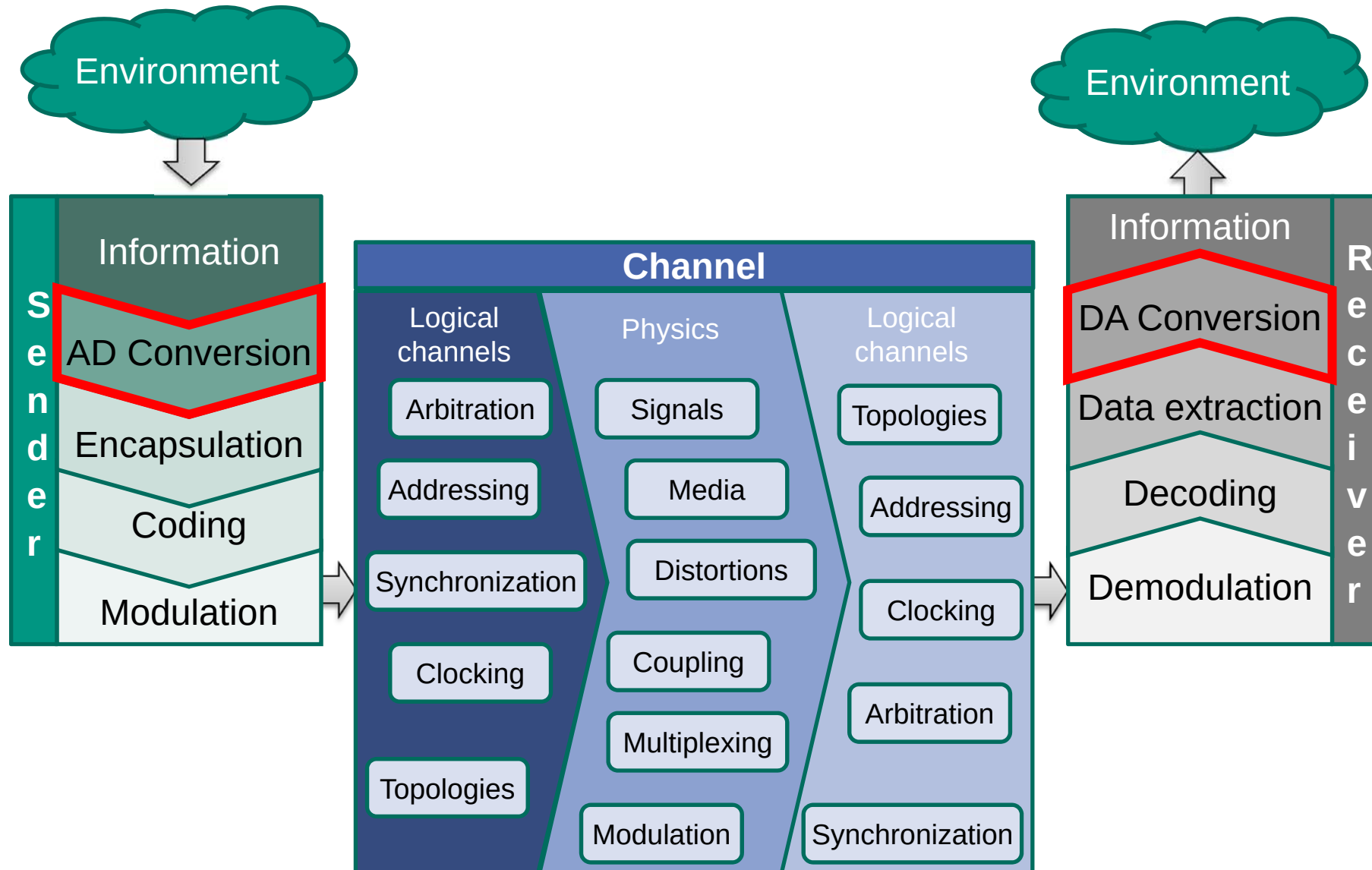
■ Capacitive



■ Couplings are DC free

- long sequences of the same value can not be detected
- a minimal frequency of signal changes must appear on medium to be transmittable into the receiving node

Transmission System – Signal Conversion

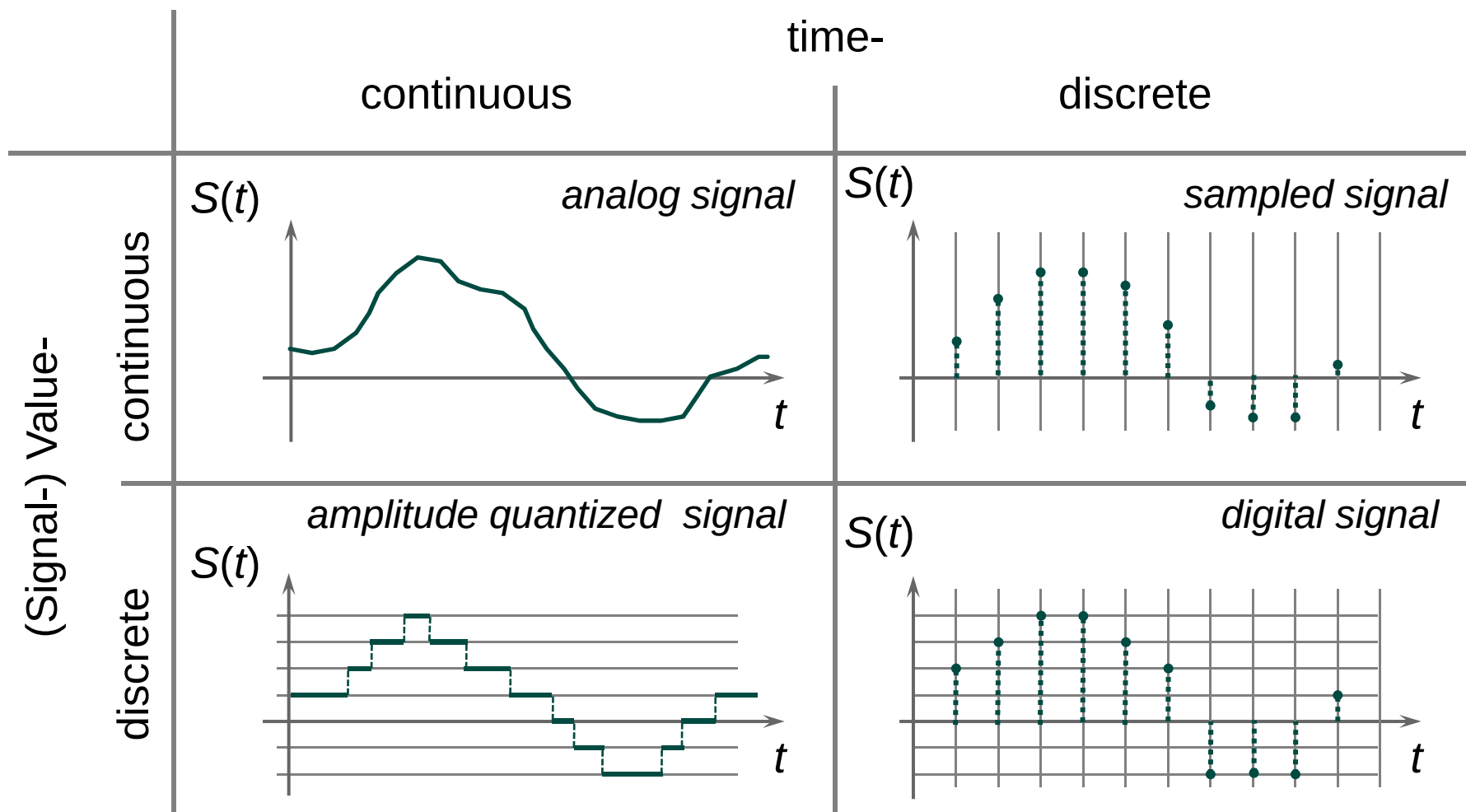


Clicker Session: Signal classes

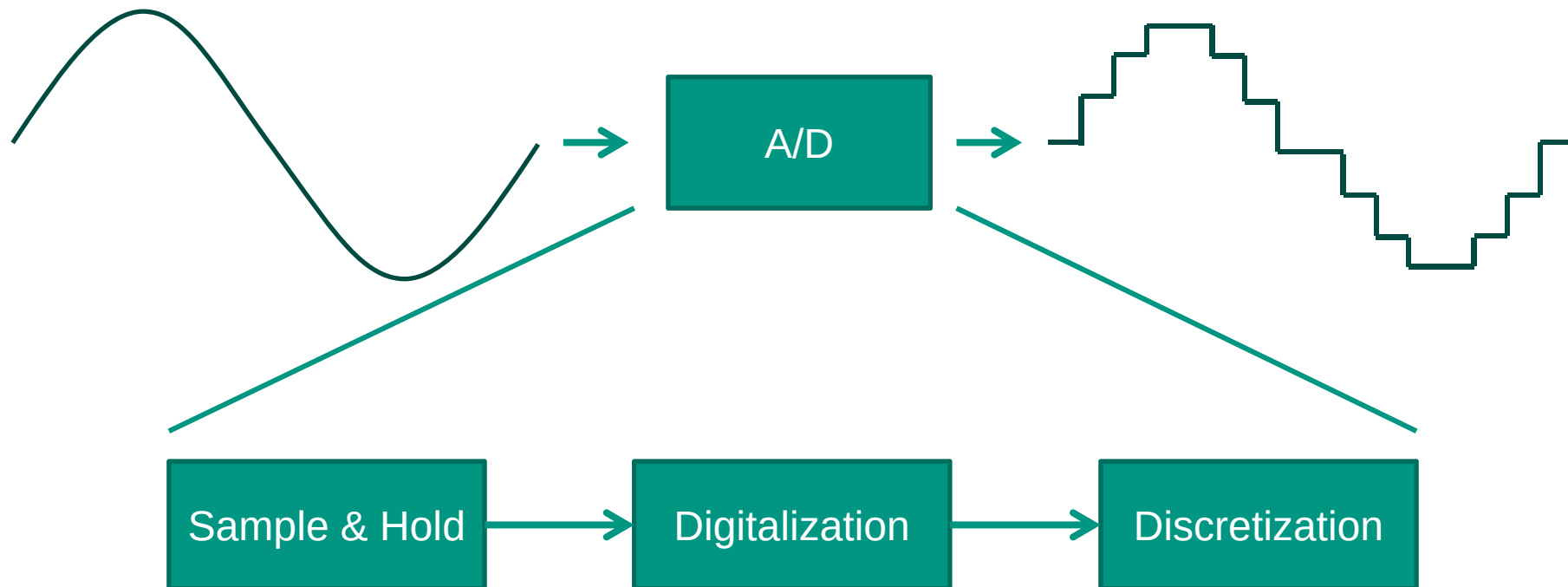
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Classes of Signals

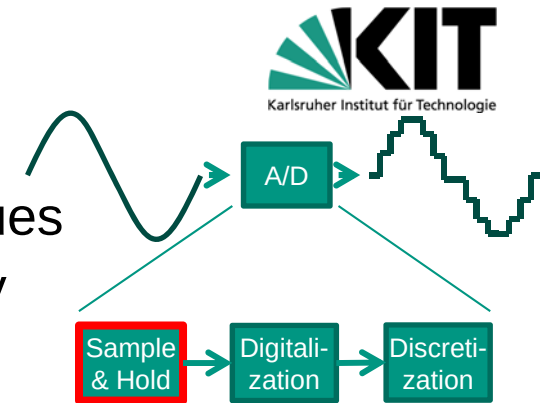


Analog-Digital Conversion



Sampling of Signals

- Transforming a continuous signal into discrete values
 - Signal shall be reconstructable unambiguously
- Prerequisites:
 - Signal limited in frequency with a maximum frequency f_{max}
 - Equidistant sampling points



■ Nyquist-Shannon-Sampling Theorem

A continuous signal has to be sampled with a frequency at least double the maximum frequency of the signal itself f_{max} in order to be able to unambiguously reconstruct the original signal.

$$f_{Sample} \geq 2 \cdot f_{max}$$

To keep this condition it may be necessary to filter the signal prior to sampling with a low-pass filter. This prevents artefacts otherwise caused by high frequencies.

Outlook for next lecture

- Circuits for sampling
- Transmission capacity
- Line codes
- Multiplexing